Numerical Study on Left-Handed Materials Made of Ferrite and Metallic Wires

UAI AIao-DIIIg(祭小共),ZHUU AIao-MIIIg(同体明),HU Geng-Kai(明史开)。

 1 Department of Applied Mechanics, Beijing Institute of Technology, Beijing 100081

²National Key Laboratory for the Prevention and Control of Explosion Disaster, Beijing Institute of Technology,

Beijing 100081

(Received 22 May 2005)

Due to coupling effect, we show that it is difficult to realize the left-handed material by placing metallic wires directly into a ferrite matrix. However by introducing an insulating material round the metallic wires to decouple the direct interaction between the metallic wire and ferrite matrix, we have proposed two microstructures, which are shown by numerical simulation to have negative refractive indexes. The in
uence of microstructure on the transmission property is also examined.

PACS: 41. 20. Jb, 78. 20. Ci, 76. 50. +g

In the past few years, left-handed material (LHM) with negative index of refraction^[1] has given rise to intense research activity, due to its potential applications in imaging^[2] and many other fields.^[3,4] The first idea of the material with both negative permittivity and negative permeability was proposed by Veselago in 1968, however only in 2001 was a lefthanded material realized for the first time by using SRR (split-ring resonance) and periodically arranged metallic wires $(WIRE)$.^[5] Since then, the research on the LHMs has been intensied, new LHM models have been proposed, $[6-8]$ and the theoretical analyses on the energy transmission and material parameter identification for the LHMs have been performed.^[9,10]

In Veselago's pioneering paper,^[1] he suggested use of conductive ferrite to provide both negative permittivity and negative permeability for the left-handed material. Many works have followed this idea to fabricate new left-handed materials.^[8,11,12] In all of these works, the metal is considered as plasma. However, no experimental or numerical computation was performed to verify these results. The essence of the WIRE arrays to have a negative permittivity is that during plasma resonance, the current induced by moving charges through the WIRE by an electric field exceeds the current induced directly by the electric field on the background, and these two currents flow in opposite direction. However when WIRE is sur rounded directly by a material with negative permeability, these two currents are of the same direction, in this case the WIRE arrays have a positive permittivity. Thus, the model of WIRE directly adding into a ferrite matrix is difficult to form an LHM. We will analyse this effect in detail through numerical simula-

We take Dewar's model (Fig. 1) $[13]$ as an example to show that the WIRE structure is hard to obtain a negative permittivity in a ferrite background. Under a static magnetic field, the ferrite has a tensorial

permeability:[11]

$$
\begin{bmatrix}\n\mu & i\kappa & 0 \\
-i\kappa & \mu & 0 \\
0 & 0 & 1\n\end{bmatrix},
$$
\n(1)

where $\mu = 1 + \frac{\gamma H_0 M_s}{2 \pi r^2}$, $\kappa = \frac{\gamma M_0}{2 \pi r^2}$ $\sqrt[2]{^2H_0^2 - \omega^2}$, $\kappa = \frac{1}{\gamma^2H_0^2 - \omega^2}$, γ is the $\gamma^2 H_0^2 - \omega^2$

gyromagnetic ratio, Ms is the saturation magnetization, H_0 is the applied static magnetic field, ω is the wave frequency. When the wave vector and the magnetic field of the TEM wave are perpendicular to H_0 , the effective relative permeability for the ferrite material can be written in the form

$$
\mu_r = (\mu^2 - \kappa^2)/\mu. \tag{2}
$$

Fig. 1. WIRE structure in ferrite matrix proposed by $Dewar.$ ^[13] The TEM wave is incident along the s direction, electrical field and static magnetic field are parallel to the WIRE.

Pendry^[14] proposed to idealize the WIRE structure as plasma, and he introduced the concept of effective mass of electrons:

$$
m_{\text{eff}} = \frac{\mu_0 \pi r^2 e^2 n}{2\pi} \ln(a/r),
$$
 (3)

Supported by the National Natural Science Foundation of China under Grant No 10325210. To whom correspondence should beaddressed. Email: hugeng@bit.edu.cn ^c 2005 Chinese Physical Society and IOP Publishing Ltd

where μ_0 is the matrix permeability, r is the WIRE radius, a is the size of unit cell, e is the electron charge, n is the active electron density in the WIRE. The plasma frequency for the WIRE structure is estimated by

$$
\omega_p^2 = n_{\text{eff}} e^2 / \varepsilon_0 m_{\text{eff}}, \qquad (4)
$$

where $n_{\text{eff}} = n \pi r / a^2$. Substituting Eq. (4) into the n_{his} is in Drude model, the effective dielectric constant of the WIRE structure is provided by

$$
\varepsilon_{\text{eff}} = 1 - \frac{\omega_p^2}{\omega(\omega + i\varepsilon_0 a^2 \omega_p^2 / \pi r^2 \sigma)},
$$
 (5)

where σ is the conductivity of the metal.

In Eq. (5), it is found that when $\omega^2 < \omega_n^2$, we have $Re(\varepsilon_{\text{eff}})$ < 0. However, when the WIRE is embedded in a ferrite matrix under a constant applied static magnetic field, in the range of negative permeability of the ferrite, the effective permittivity of the WIRE may not be negative due to the coupling effect. Back to formula (3), due to the negative permeability of the ferrite matrix, the effective mass of electrons has the

$$
m_{\text{eff}} = \frac{\mu_r \mu_0 \pi r^2 e^2 n}{2\pi} \ln(a/r) < 0,\tag{6}
$$

where μ_r is the relative permeability of the ferrite, and $\mu_r < 0$. Substituting Eq. (6) into Eqs. (4) and (5), we observe that $\omega_p^2 < 0$ (the complex frequency has been discussed by Wu et a_i . \cdots is so no matter whether \cdots ω^2 < $|\omega_p^2|$ or $\omega^2 > |\omega_p^2|$, we have Re(ε_{eff}) > 0.

It can be concluded that if the WIRE is embedded into a matrix with a negative permeability, due to the coupling effect, it is difficult to obtain a negative effective permittivity of the composite. $Dewar^{[8]}$ arrived at the same conclusion by solving a boundary value problem.

Fig. 2. Transmission property of WIRE, ferrite and their composite. Here $\omega_m = \gamma \sqrt{H_0(H_0 + M_s)}$ is the ferromagnetic resonance frequency, and $\omega_{mp} = \gamma (H_0 + M_s)$ is the anti-resonance frequency.

In addition, as ferrite usually possesses high relative permittivity, $\varepsilon_r \varepsilon_0$ should be utilized to replace ε_0 in Eq. (4) . This significantly reduces the plasma frequency of the WIRE, typically to one tenth of that in vacuum.

 $\tau/\varepsilon_0 m_{\text{eff}}$, and τ_4 internal measure permeability is higher than the range Consequently, the frequency range where the fernecessary for a negative permittivity of the WIRE. This is illustrated in Fig. 2, due to the coupling effect, usually there is no overlap in the frequency for both negative permittivity and permeability for the com posite, necessary to form a left-handed material.

 $\sqrt{7}$ From the above analysis, the key point to make \langle (b) direction. From the point of fabrication, models B LHM from ferrite and WIRE is to decouple this direct electromagnetic interaction. To this end, the following three modications can be proposed to reduce this interaction, as shown in Fig. 3. Model A (proposed first by Dewar^[8]) is that the WIRE is surrounded by an insulating layer, and the whole cell is together placed into a ferrite matrix. Model B is that the WIRE is placed into an insulating matrix, they are together laid with ferrite vertically, model C is that the WIRE is placed in an insulating matrix, and then they are together laid with ferrite horizontally. In all these three models, a static magnetic field is applied along the Y and C are more easy to manipulate than model A . Model C is similar to SRR plus a WIRE structure as proposed by Shelby ev al., where the SRR is replaced by the ferrite material. In the following, we analyse the transmission property of these materials.

Fig. 3. Top view of models to reduce electromagnetic coupling. The TEM wave is incident along the z direction, with magnetic field parallel to x , the electric field and applied static magnetic field parallel to y .

Finite element calculation on a ferrite material with known property is first performed to check the numerical method. Consider a ferrite of 3p-6-15 mm, which lies in a parallel plate waveguide $(Fig. 4)$, where p is the number of the element repeated in the x direction. The waveguide is a vacuum and two sides of the x direction are set to be master-slave boundary condition (periodic condition) with zero phase delay. Of all

models in this study, the ferrite is supposed to have a saturation magnetization of 1700 G, and a relative dielectric constant 13. The applied static magnetic field is 1256 Oe.

When a TEM wave is incident along the z direction, with magnetic field parallel to the x direction, and electric field parallel to the y direction, substituting ferrite's tensorial permeability (Eq. (1)) into plane wave propagation equation, and noting $k = k_x$, then S-parameters can be estimated analytically, and they are given by the following form:[10]

$$
S_{11} = \frac{R(1 - T^2)}{1 - R^2 T^2},
$$

\n
$$
S_{21} = \frac{T(1 - R^2)}{1 - R^2 T^2},
$$
\n(7)

where $R = (z - 1)/(z + 1);$ $T = \exp(-i\sqrt{\varepsilon_r \mu_r} k_0 d);$ μ_r is the ferrite effective permeability given by Eq. (2) ; k_0 is the vacuum wave vector; d is the thickness of model. Figure 5 gives the S-parameter S_{21} estimated from Eq. (7) and calculated by the HFSS method.^[15]

Fig. 4. Finite element model for a plate waveguide, periodic condition along the x direction. Top and bottom surfaces in the y direction are PEC, front and back surfaces in the z direction are two waveports.

Fig. 5. Comparison between analytical results (dotted line) and HFSS result (solid line) for transmission property S_{21} .

As shown in Fig. 5, both the analytical and HFSS results for the S_{21} parameter predict stopband extending from 5.3 to 8.2 GHz, in which S_{21} is less than -30 dB (see Fig. 5). The results predicted by the two methods agree well in a large part of frequency range.

The curve rises less shapely at the right part within the stopband, this is due to a small negative permeability of the ferrite at these frequencies. The applied static magnetic field also interferes with the stopband, which is explained in Fig. 2.

 γ is within the stopbands of the ferrite matrix, however The finite element model for the new proposed structures proposed previously is the same as shown in Fig. 4. The transmission property S_{21} is calculated for the three microstructures, the computed results are shown in Fig. 6, the result of Dewar's initial model (Fig. 1) is also included for comparison. Compared to the pure ferrite material (Fig. 5) and Dewar's initial model, it is found that there exist some passbands for the composites (models A, B, C) approximately spanning from 6.75 to 7.75 GHz, originally these passbands the initial model (Fig. 1) proposed by Dewar has no passband. This indicates that the proposed composites have both negative permittivity and permeability. We have checked from the phase diagram, indeed in this frequency range, the composites are left-handed materials. In the three models, the ferrite occupies a volume function of 66%, 73% and 67%, respectively. The radii of all metallic wires are 0.15 mm.

> The simulated results show that by including an insulating material round metallic wires, the electromagnetic coupling effect is greatly reduced. This makes left-handed materials possible from ferrite and WIRE. According to Eq. (4), the presence of large vol ume fraction of ferrite lowers the plasma frequency of the WIRE typically from 50 to 7 GHz, this is why S_{21} is still very high for the frequency above 8.2 GHz. When the frequency is slightly larger than the reso nance frequency of the ferrite (5.3 GHz), the absolute value of the ferrite permeability is reduced quickly, the effective permeability of the element is dominated by the WIRE and insulating material.

 \mathbf{r} ig. \mathbf{o} . Simulated transmission property \mathcal{S}_{21} for models A, B, and C, and Dewar's initial model.

Figure 7 shows the electric field variation as a function of time along the central line (along the z direction) of the models A , B and C , respectively. For the electric field $E_0 \exp(i(kz-wt))$, let the phase be a constant (here we take to be 0), we can obtain constant phase diagram, which can be written as $[16]$

$$
z = \frac{1}{k}\omega t. \tag{8}
$$

The slopes of the black line in Fig. 7 are negative, indicating that their wave vector k is negative. Refractive indexes of the composites can be estimated directly from Fig. 7 by k , these results are listed in Table 1, the complex refractive indexes can also be evaluated

directly from transmission property S_{11} and S_{21} , the \blacksquare they are also listed for comparison. A good agreement

 k , and the minimal design for the minimal design for the minimal design for the minimal design for the minimal design ℓ crostructure, the influence of microstructure parameters on the transmission property is examined, the computed S_{21} as a function of the ferrite volume fraction and the radius of WIRE for model C at 7.0 GHz is shown in Fig. 8. In the computation, the WIRE is

Fig. 7. Electric eld variation through the centre of models A, B, and C, at frequencies of 7.0, 7.2 and 7.0 GHz, respectively.

Fig. 8. Inhuence of ferrite volume fraction and WIRE α C_{mean} radius on transmission property S_{21} .

assumed to be placed in the middle of insulating ma terial. When the radius of the WIRE varies from 0.10 [10] to 0.24 mm , S_{21} is only slightly reduced. It is also found that S_{21} reaches a maximum value at 60% volume fraction of ferrite (the 1.8 mm width of the ferrite in the cell). Thus optimal design of the microstructure for the left-handed material will be useful, and this will be a subject of our future works.

In conclusion, we have shown that due to the electromagnetic coupling effect, it is difficult to make lefthanded materials directly by embedding metallic wires into ferrite matrix. By introducing an insulating ma terial round the WIRE, we have proposed two new microstructures made of WIRE, insulating material and ferrite, which are shown by numerical simulation to have negative refractive indices in certain frequency.

References

- Veselago V G 1968 Sov. Phys. Usp. $10\ 509$ $\lceil 1 \rceil$
- Pendry J B 2000 Phys. Rev. Lett. 85 3966
- [3] Schwartz B T and Piestun R 2004 Appl. Phys. Lett. 85 1
- [4] Li J, Zhou L, Chan C T and Sheng P 2003 Phys. Rev. Lett. 908 $-$
- $\mathsf{I}5$ Shelby R A, Smith D R and Schultz S 2001 Science 292 77
- [6] Sanada A, Caloz C and Tatsuo I 2004 IEEE T. Microwave Theory ⁵² 1252
- [7] Chen H S, Ran L X, Huangfu J T, Zhang X M, Chen K S, Grzegorezyk T M and Kong J A 2004 Phys. Rev. E 70 057605
- [8] Dewar G 2003 Proc. SPIE ⁵²¹⁸ 140
- Cui T J and Kong J A 2004 Phys. Rev. B 70 205106
- Chen X D, Grzegorezyk T M, Wu B L, Pacheco J and Kong J A 2004 Phys. Rev. E ⁷⁰ 016608
- [11] Chui S T and Hu L B 2000 Phys. Rev. B ⁶⁵ 144407
- [12] Wu R X 2005 J. Appl. Phys. ⁹⁷ 076105
- [13] Dewar G 2000 Int. J. Mod. Phys. B ¹⁵ 3258
- [14] Pendry J B 1996 Phys. Rev. Lett. 76 4773
- [15] Ansoft HFSS (High Frequency Structure Simulator) http://www.ansoft.com.cn
- [16] Moss C D, Grzegorezyk T M, Zhang Y and Kong J A 2002 Prog. Electromagn. Res. ³⁵ 315
- [17] Wu L, He S and Shen L 2003 Phys. Rev. B ⁶⁷ 235103

射 频 和 天 线 设 计 培 训 课 程 推 荐

易迪拓培训(www.edatop.com)由数名来自于研发第一线的资深工程师发起成立,致力并专注于微 波、射频、天线设计研发人才的培养;我们于 2006 年整合合并微波 EDA 网(www.mweda.com), 现 已发展成为国内最大的微波射频和天线设计人才培养基地,成功推出多套微波射频以及天线设计经典 培训课程和 ADS、HFSS 等专业软件使用培训课程,广受客户好评;并先后与人民邮电出版社、电子 工业出版社合作出版了多本专业图书,帮助数万名工程师提升了专业技术能力。客户遍布中兴通讯、 研通高频、埃威航电、国人通信等多家国内知名公司,以及台湾工业技术研究院、永业科技、全一电 子等多家台湾地区企业。

易迪拓培训课程列表:http://www.edatop.com/peixun/rfe/129.html

射频工程师养成培训课程套装

该套装精选了射频专业基础培训课程、射频仿真设计培训课程和射频电 路测量培训课程三个类别共30门视频培训课程和3本图书教材;旨在 引领学员全面学习一个射频工程师需要熟悉、理解和掌握的专业知识和 研发设计能力。通过套装的学习,能够让学员完全达到和胜任一个合格 的射频工程师的要求…

课程网址:http://www.edatop.com/peixun/rfe/110.html

ADS 学习培训课程套装

该套装是迄今国内最全面、最权威的 ADS 培训教程, 共包含 10 门 ADS 学习培训课程。课程是由具有多年 ADS 使用经验的微波射频与通信系 统设计领域资深专家讲解,并多结合设计实例,由浅入深、详细而又 全面地讲解了 ADS 在微波射频电路设计、通信系统设计和电磁仿真设 计方面的内容。能让您在最短的时间内学会使用 ADS,迅速提升个人技 术能力, 把 ADS 真正应用到实际研发工作中去, 成为 ADS 设计专家...

课程网址: http://www.edatop.com/peixun/ads/13.html

HFSS 学习培训课程套装

该套课程套装包含了本站全部 HFSS 培训课程,是迄今国内最全面、最 专业的HFSS培训教程套装,可以帮助您从零开始,全面深入学习HFSS 的各项功能和在多个方面的工程应用。购买套装,更可超值赠送 3 个月 免费学习答疑,随时解答您学习过程中遇到的棘手问题,让您的 HFSS 学习更加轻松顺畅…

课程网址:http://www.edatop.com/peixun/hfss/11.html

CST 学习培训课程套装

该培训套装由易迪拓培训联合微波 EDA 网共同推出, 是最全面、系统、 专业的 CST 微波工作室培训课程套装,所有课程都由经验丰富的专家授 课, 视频教学, 可以帮助您从零开始, 全面系统地学习 CST 微波工作的 各项功能及其在微波射频、天线设计等领域的设计应用。且购买该套装, 还可超值赠送 3 个月免费学习答疑…

课程网址:http://www.edatop.com/peixun/cst/24.html

HFSS 天线设计培训课程套装

套装包含 6 门视频课程和 1 本图书, 课程从基础讲起, 内容由浅入深, 理论介绍和实际操作讲解相结合,全面系统的讲解了 HFSS 天线设计的 全过程。是国内最全面、最专业的 HFSS 天线设计课程,可以帮助您快 速学习掌握如何使用 HFSS 设计天线,让天线设计不再难…

课程网址:http://www.edatop.com/peixun/hfss/122.html

13.56MHz NFC/RFID 线圈天线设计培训课程套装

套装包含 4 门视频培训课程, 培训将 13.56MHz 线圈天线设计原理和仿 真设计实践相结合, 全面系统地讲解了13.56MHz 线圈天线的工作原理、 设计方法、设计考量以及使用 HFSS 和 CST 仿真分析线圈天线的具体 操作,同时还介绍了13.56MHz 线圈天线匹配电路的设计和调试。通过 该套课程的学习,可以帮助您快速学习掌握 13.56MHz 线圈天线及其匹 配电路的原理、设计和调试…

详情浏览: http://www.edatop.com/peixun/antenna/116.html

我们的课程优势:

- ※ 成立于 2004 年,10 多年丰富的行业经验,
- ※ 一直致力并专注于微波射频和天线设计工程师的培养,更了解该行业对人才的要求
- ※ 经验丰富的一线资深工程师讲授,结合实际工程案例,直观、实用、易学

联系我们:

- ※ 易迪拓培训官网: http://www.edatop.com
- ※ 微波 EDA 网: http://www.mweda.com
- ※ 官方淘宝店: http://shop36920890.taobao.com

专注于微波、射频、天线设计人才的培养

男油拓語训 官方网址: http://www.edatop.com 淘宝网店:http://shop36920890.taobao.com